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Digital cellular telecommunications system (Phase 2+) (GSM); Enhanced Full Rate (EFR) speech transcoding (GSM 06.60 version 8.0.1 Release 1999)

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**ICS:**

33.070.50	Globalni sistem za mobilno telekomunikacijo (GSM)	Global System for Mobile Communication (GSM)
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# ETSI EN 300 726 V8.0.1 (2000-11)

European Standard (Telecommunications series)

**Digital cellular telecommunications system (Phase 2+);  
Enhanced Full Rate (EFR) speech transcoding;  
(GSM 06.60 version 8.0.1 Release 1999)**

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**ETSI**

650 Route des Lucioles  
F-06921 Sophia Antipolis Cedex - FRANCE

Tel.: +33 4 92 94 42 00 Fax: +33 4 93 65 47 16

Siret N° 348 623 562 00017 - NAF 742 C  
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## Foreword

This European Standard (Telecommunications series) has been produced by ETSI Technical Committee Special Mobile Group (SMG).

The present document describes the detailed mapping between input blocks of 160 speech samples in 13-bit uniform PCM format to encoded blocks of 244 bits and from encoded blocks of 244 bits to output blocks of 160 reconstructed speech samples within the digital cellular telecommunications system.

The contents of the present document is subject to continuing work within SMG and may change following formal SMG approval. Should SMG modify the contents of the present document it will be re-released with an identifying change of release date and an increase in version number as follows:

Version 8.x.y

where:

- 8 indicates Release 1999 of GSM Phase 2+
- x the second digit is incremented for all changes of substance, i.e. technical enhancements, corrections, updates, etc.
- y the third digit is incremented when editorial only changes have been incorporated in the specification.

### National transposition dates

Date of adoption of this EN:	3 November 2000
Date of latest announcement of this EN (doa):	28 February 2001
Date of latest publication of new National Standard or endorsement of this EN (dop/e):	31 August 2001
Date of withdrawal of any conflicting National Standard (dow):	31 August 2001

# 1 Scope

The present document describes the detailed mapping between input blocks of 160 speech samples in 13-bit uniform PCM format to encoded blocks of 244 bits and from encoded blocks of 244 bits to output blocks of 160 reconstructed speech samples. The sampling rate is 8 000 sample/s leading to a bit rate for the encoded bit stream of 12,2 kbit/s. The coding scheme is the so-called Algebraic Code Excited Linear Prediction Coder, hereafter referred to as ACELP.

The present document also specifies the conversion between A-law or  $\mu$ -law (PCS 1900) PCM and 13-bit uniform PCM. Performance requirements for the audio input and output parts are included only to the extent that they affect the transcoder performance. This part also describes the codec down to the bit level, thus enabling the verification of compliance to the part to a high degree of confidence by use of a set of digital test sequences. These test sequences are described in GSM 06.54 [7] and are available on disks.

In case of discrepancy between the requirements described in the present document and the fixed point computational description (ANSI-C code) of these requirements contained in GSM 06.53 [6], the description in GSM 06.53 [6] will prevail.

The transcoding procedure specified in the present document is applicable for the enhanced full rate speech traffic channel (TCH) in the GSM system.

In GSM 06.51 [5], a reference configuration for the speech transmission chain of the GSM enhanced full rate (EFR) system is shown. According to this reference configuration, the speech encoder takes its input as a 13-bit uniform PCM signal either from the audio part of the Mobile Station or on the network side, from the PSTN via an 8-bit/A-law or  $\mu$ -law (PCS 1900) to 13-bit uniform PCM conversion. The encoded speech at the output of the speech encoder is delivered to a channel encoder unit which is specified in GSM 05.03 [3]. In the receive direction, the inverse operations take place.

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# 2 References

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The following documents contain provisions which, through reference in this text, constitute provisions of the present document.

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- References are either specific (identified by date of publication, edition number, version number, etc.) or non-specific.
- For a specific reference, subsequent revisions do not apply.
- For a non-specific reference, the latest version applies.
- A non-specific reference to an ETS shall also be taken to refer to later versions published as an EN with the same number.
- For this Release 1999 document, references to GSM documents are for Release 1999 versions (version 8.x.y).

- [1] GSM 01.04: "Digital cellular telecommunications system (Phase 2+); Abbreviations and acronyms".
- [2] GSM 03.50: "Digital cellular telecommunications system (Phase 2+); Transmission planning aspects of the speech service in the GSM Public Land Mobile Network (PLMN) system".
- [3] GSM 05.03: "Digital cellular telecommunications system (Phase 2+); Channel coding".
- [4] GSM 06.32: "Digital cellular telecommunications system (Phase 2+); Voice Activity Detection (VAD)".
- [5] GSM 06.51: "Digital cellular telecommunications system (Phase 2+); Enhanced Full Rate (EFR) speech processing functions General description".
- [6] GSM 06.53: "Digital cellular telecommunications system (Phase 2+); ANSI-C code for the GSM Enhanced Full Rate (EFR) speech codec".

- [7] GSM 06.54: "Digital cellular telecommunications system (Phase 2+); Test vectors for the GSM Enhanced Full Rate (EFR) speech codec".
- [8] ITU-T Recommendation G.711 (1988): "Coding of analogue signals by pulse code modulation Pulse code modulation (PCM) of voice frequencies".
- [9] ITU-T Recommendation G.726: "40, 32, 24, 16 kbit/s adaptive differential pulse code modulation (ADPCM)".

## 3 Definitions, symbols and abbreviations

### 3.1 Definitions

For the purposes of the present document, the following terms and definitions apply:

**adaptive codebook:** adaptive codebook contains excitation vectors that are adapted for every subframe. The adaptive codebook is derived from the long term filter state. The lag value can be viewed as an index into the adaptive codebook.

**adaptive postfilter:** this filter is applied to the output of the short term synthesis filter to enhance the perceptual quality of the reconstructed speech. In the GSM enhanced full rate codec, the adaptive postfilter is a cascade of two filters: a formant postfilter and a tilt compensation filter.

**algebraic codebook:** fixed codebook where algebraic code is used to populate the excitation vectors (innovation vectors). The excitation contains a small number of nonzero pulses with predefined interlaced sets of positions.

**closed-loop pitch analysis:** this is the adaptive codebook search, i.e., a process of estimating the pitch (lag) value from the weighted input speech and the long term filter state. In the closed-loop search, the lag is searched using error minimization loop (analysis-by-synthesis). In the GSM enhanced full rate codec, closed-loop pitch search is performed for every subframe.

**direct form coefficients:** one of the formats for storing the short term filter parameters. In the GSM enhanced full rate codec, all filters which are used to modify speech samples use direct form coefficients.

**fixed codebook:** fixed codebook contains excitation vectors for speech synthesis filters. The contents of the codebook are non-adaptive (i.e., fixed). In the GSM enhanced full rate codec, the fixed codebook is implemented using an algebraic codebook.

**fractional lags:** set of lag values having sub-sample resolution. In the GSM enhanced full rate codec a sub-sample resolution of 1/6th of a sample is used.

**frame:** time interval equal to 20 ms (160 samples at an 8 kHz sampling rate).

**integer lags:** set of lag values having whole sample resolution.

**interpolating filter:** FIR filter used to produce an estimate of sub-sample resolution samples, given an input sampled with integer sample resolution.

**inverse filter:** this filter removes the short term correlation from the speech signal. The filter models an inverse frequency response of the vocal tract.

**lag:** long term filter delay. This is typically the true pitch period, or a multiple or sub-multiple of it.

**Line Spectral Frequencies:** (see Line Spectral Pair).

**Line Spectral Pair:** transformation of LPC parameters. Line Spectral Pairs are obtained by decomposing the inverse filter transfer function  $A(z)$  to a set of two transfer functions, one having even symmetry and the other having odd symmetry. The Line Spectral Pairs (also called as Line Spectral Frequencies) are the roots of these polynomials on the z-unit circle).



**LP analysis window:** for each frame, the short term filter coefficients are computed using the high pass filtered speech samples within the analysis window. In the GSM enhanced full rate codec, the length of the analysis window is 240 samples. For each frame, two asymmetric windows are used to generate two sets of LP coefficients. No samples of the future frames are used (no lookahead).

**LP coefficients:** Linear Prediction (LP) coefficients (also referred as Linear Predictive Coding (LPC) coefficients) is a generic descriptive term for describing the short term filter coefficients.

**open-loop pitch search:** process of estimating the near optimal lag directly from the weighted speech input. This is done to simplify the pitch analysis and confine the closed-loop pitch search to a small number of lags around the open-loop estimated lags. In the GSM enhanced full rate codec, open-loop pitch search is performed every 10 ms.

**residual:** output signal resulting from an inverse filtering operation.

**short term synthesis filter:** this filter introduces, into the excitation signal, short term correlation which models the impulse response of the vocal tract.

**perceptual weighting filter:** this filter is employed in the analysis-by-synthesis search of the codebooks. The filter exploits the noise masking properties of the formants (vocal tract resonances) by weighting the error less in regions near the formant frequencies and more in regions away from them.

**subframe:** time interval equal to 5 ms (40 samples at an 8 kHz sampling rate).

**vector quantization:** method of grouping several parameters into a vector and quantizing them simultaneously.

**zero input response:** output of a filter due to past inputs, i.e. due to the present state of the filter, given that an input of zeros is applied.

**zero state response:** output of a filter due to the present input, given that no past inputs have been applied, i.e., given the state information in the filter is all zeroes.

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## 3.2 Symbols

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For the purposes of the present document, the following symbols apply:

$A(z)$	The inverse filter with unquantized coefficients
$\hat{A}(z)$	The inverse filter with quantified coefficients
$H(z) = \frac{1}{\hat{A}(z)}$	The speech synthesis filter with quantified coefficients
$a_i$	The unquantized linear prediction parameters (direct form coefficients)
$\hat{a}_i$	The quantified linear prediction parameters
$m$	The order of the LP model
$\frac{1}{B(z)}$	The long-term synthesis filter
$W(z)$	The perceptual weighting filter (unquantized coefficients)
$\gamma_1, \gamma_2$	The perceptual weighting factors
$F_E(z)$	Adaptive pre-filter
$T$	The nearest integer pitch lag to the closed-loop fractional pitch lag of the subframe
$\beta$	The adaptive pre-filter coefficient (the quantified pitch gain)
$H_f(z) = \frac{\hat{A}(z / \gamma_n)}{\hat{A}(z / \gamma_d)}$	The formant postfilter
$\gamma_n$	Control coefficient for the amount of the formant post-filtering
$\gamma_d$	Control coefficient for the amount of the formant post-filtering
$H_t(z)$	Tilt compensation filter

$\gamma_t$	Control coefficient for the amount of the tilt compensation filtering
$\mu = \gamma_t k_1'$	A tilt factor, with $k_1'$ being the first reflection coefficient
$h_f(n)$	The truncated impulse response of the formant postfilter
$L_h$	The length of $h_f(n)$
$r_h(i)$	The auto-correlations of $h_f(n)$
$\hat{A}(z/\gamma_n)$	The inverse filter (numerator) part of the formant postfilter
$1/\hat{A}(z/\gamma_d)$	The synthesis filter (denominator) part of the formant postfilter
$\hat{r}(n)$	The residual signal of the inverse filter $\hat{A}(z/\gamma_n)$
$h_t(z)$	Impulse response of the tilt compensation filter
$\beta_{sc}(n)$	The AGC-controlled gain scaling factor of the adaptive postfilter
$\alpha$	The AGC factor of the adaptive postfilter
$H_{hl}(z)$	Pre-processing high-pass filter
$w_I(n), w_{II}(n)$	LP analysis windows
$L_1^{(I)}$	Length of the first part of the LP analysis window $w_I(n)$
$L_2^{(I)}$	Length of the second part of the LP analysis window $w_I(n)$
$L_1^{(II)}$	Length of the first part of the LP analysis window $w_{II}(n)$
$L_2^{(II)}$	Length of the second part of the LP analysis window $w_{II}(n)$
$r_{ac}(k)$	The auto-correlations of the windowed speech $s'(n)$
$w_{lag}(i)$	Lag window for the auto-correlations (60 Hz bandwidth expansion)
$f_0$	The bandwidth expansion in Hz
$f_s$	The sampling frequency in Hz
$r'_{ac}(k)$	The modified (bandwidth expanded) auto-correlations
$E_{LD}(i)$	The prediction error in the $i$ th iteration of the Levinson algorithm
$k_i$	The $i$ th reflection coefficient
$a_j^{(i)}$	The $j$ th direct form coefficient in the $i$ th iteration of the Levinson algorithm
$F_1'(z)$	Symmetric LSF polynomial
$F_2'(z)$	Antisymmetric LSF polynomial
$F_1(z)$	Polynomial $F_1'(z)$ with root $z = -1$ eliminated
$F_2(z)$	Polynomial $F_2'(z)$ with root $z = 1$ eliminated
$q_i$	The line spectral pairs (LSPs) in the cosine domain
$\mathbf{q}$	An LSP vector in the cosine domain
$\hat{\mathbf{q}}_i^{(n)}$	The quantified LSP vector at the $i$ th subframe of the frame $n$
$\omega_i$	The line spectral frequencies (LSFs)
$T_m(x)$	A $m$ th order Chebyshev polynomial
$f_1(i), f_2(i)$	The coefficients of the polynomials $F_1(z)$ and $F_2(z)$
$f_1'(i), f_2'(i)$	The coefficients of the polynomials $F_1'(z)$ and $F_2'(z)$
$f(i)$	The coefficients of either $F_1(z)$ or $F_2(z)$
$C(x)$	Sum polynomial of the Chebyshev polynomials
$x$	Cosine of angular frequency $\omega$

$\lambda_k$	Recursion coefficients for the Chebyshev polynomial evaluation
$f_i$	The line spectral frequencies (LSFs) in Hz
$\mathbf{f}^t = [f_1 f_2 \dots f_{10}]$	The vector representation of the LSFs in Hz
$\mathbf{z}^{(1)}(n), \mathbf{z}^{(2)}(n)$	The mean-removed LSF vectors at frame $n$
$\mathbf{r}^{(1)}(n), \mathbf{r}^{(2)}(n)$	The LSF prediction residual vectors at frame $n$
$\mathbf{p}(n)$	The predicted LSF vector at frame $n$
$\hat{\mathbf{r}}^{(2)}(n-1)$	The quantified second residual vector at the past frame
$\hat{\mathbf{f}}^k$	The quantified LSF vector at quantization index $k$
$E_{LSP}$	The LSP quantization error
$w_i, i = 1, \dots, 10$	LSP-quantization weighting factors
$d_i$	The distance between the line spectral frequencies $f_{i+1}$ and $f_{i-1}$
$h(n)$	The impulse response of the weighted synthesis filter
$O_k$	The correlation maximum of open-loop pitch analysis at delay $k$
$O_i, i = 1, \dots, 3$	The correlation maxima at delays $t_i, i = 1, \dots, 3$
$(M_i, t_i), i = 1, \dots, 3$	The normalized correlation maxima $M_i$ and the corresponding delays $t_i, i = 1, \dots, 3$
$H(z)W(z) = \frac{A(z/\gamma_1)}{\hat{A}(z)A(z/\gamma_2)}$	The weighted synthesis filter
$A(z/\gamma_1)$	The numerator of the perceptual weighting filter
$1/A(z/\gamma_2)$	The denominator of the perceptual weighting filter
$T_1$	The nearest integer to the fractional pitch lag of the previous (1st or 3rd) subframe
$s'(n)$	The windowed speech signal
$s_w(n)$	The weighted speech signal
$\hat{s}(n)$	Reconstructed speech signal
$\hat{s}'(n)$	The gain-scaled post-filtered signal
$\hat{s}_f(n)$	Post-filtered speech signal (before scaling)
$x(n)$	The target signal for adaptive codebook search
$x_2(n), \mathbf{x}_2^t$	The target signal for algebraic codebook search
$res_{LP}(n)$	The LP residual signal
$c(n)$	The fixed codebook vector
$v(n)$	The adaptive codebook vector
$y(n) = v(n) * h(n)$	The filtered adaptive codebook vector
$y_k(n)$	The past filtered excitation
$u(n)$	The excitation signal
$\hat{u}(n)$	The emphasized adaptive codebook vector
$\hat{u}'(n)$	The gain-scaled emphasized excitation signal
$T_{op}$	The best open-loop lag
$t_{min}$	Minimum lag search value
$t_{max}$	Maximum lag search value
$R(k)$	Correlation term to be maximized in the adaptive codebook search
$b_{24}$	The FIR filter for interpolating the normalized correlation term $R(k)$
$R(k)_t$	The interpolated value of $R(k)$ for the integer delay $k$ and fraction $t$

$b_{60}$	The FIR filter for interpolating the past excitation signal $u(n)$ to yield the adaptive codebook vector $v(n)$
$A_k$	Correlation term to be maximized in the algebraic codebook search at index $k$
$C_k$	The correlation in the numerator of $A_k$ at index $k$
$E_{Dk}$	The energy in the denominator of $A_k$ at index $k$
$\mathbf{d} = \mathbf{H}^t \mathbf{x}_2$	The correlation between the target signal $x_2(n)$ and the impulse response $h(n)$ , i.e., backward filtered target
$\mathbf{H}$	The lower triangular Toeplitz convolution matrix with diagonal $h(0)$ and lower diagonals $h(1), \dots, h(39)$
$\Phi = \mathbf{H}^t \mathbf{H}$	The matrix of correlations of $h(n)$
$d(n)$	The elements of the vector $\mathbf{d}$
$\phi(i, j)$	The elements of the symmetric matrix $\Phi$
$\mathbf{c}_k$	The innovation vector
$C$	The correlation in the numerator of $A_k$
$m_i$	The position of the $i$ th pulse
$\vartheta_i$	The amplitude of the $i$ th pulse
$N_p$	The number of pulses in the fixed codebook excitation
$E_D$	The energy in the denominator of $A_k$
$res_{LTP}(n)$	The normalized long-term prediction residual
$b(n)$	The sum of the normalized $d(n)$ vector and normalized long-term prediction residual $res_{LTP}(n)$
$s_b(n)$	The sign signal for the algebraic codebook search
$d'(n)$	Sign extended backward filtered target
$\phi'(i, j)$	The modified elements of the matrix $\Phi$ , including sign information
$\mathbf{z}^t, z(n)$	The fixed codebook vector convolved with $h(n)$
$E(n)$	The mean-removed innovation energy (in dB)
$\bar{E}$	The mean of the innovation energy
$\tilde{E}(n)$	The predicted energy
$[b_1 \ b_2 \ b_3 \ b_4]$	The MA prediction coefficients
$\hat{R}(k)$	The quantified prediction error at subframe $k$
$E_I$	The mean innovation energy
$R(n)$	The prediction error of the fixed-codebook gain quantization
$E_Q$	The quantization error of the fixed-codebook gain quantization
$e(n)$	The states of the synthesis filter $1/\hat{A}(z)$
$e_w(n)$	The perceptually weighted error of the analysis-by-synthesis search
$\eta$	The gain scaling factor for the emphasized excitation
$g_c$	The fixed-codebook gain
$g_c'$	The predicted fixed-codebook gain
$\hat{g}_c$	The quantified fixed codebook gain
$g_p$	The adaptive codebook gain
$\hat{g}_p$	The quantified adaptive codebook gain

$\gamma_{gc} = g_c / g'_c$	A correction factor between the gain $g_c$ and the estimated one $g'_c$
$\hat{\gamma}_{gc}$	The optimum value for $\gamma_{gc}$
$\gamma_{sc}$	Gain scaling factor

### 3.3 Abbreviations

For the purposes of the present document, the following abbreviations apply. Further GSM related abbreviations may be found in GSM 01.04 [1].

ACELP	Algebraic Code Excited Linear Prediction
AGC	Adaptive Gain Control
CELP	Code Excited Linear Prediction
FIR	Finite Impulse Response
ISPP	Interleaved Single-Pulse Permutation
LP	Linear Prediction
LPC	Linear Predictive Coding
LSF	Line Spectral Frequency
LSP	Line Spectral Pair
LTP	Long Term Predictor (or Long Term Prediction)
MA	Moving Average

## 4 Outline description

The present document is structured as follows.

Subclause 4.1 contains a functional description of the audio parts including the A/D and D/A functions. Subclause 4.2 describes the conversion between 13-bit uniform and 8-bit A-law or  $\mu$ -law (PCS 1900) samples. Subclauses 4.3 and 4.4 present a simplified description of the principles of the GSM EFR encoding and decoding process respectively. In clause 4.5, the sequence and subjective importance of encoded parameters are given.

Clause 5 presents the functional description of the GSM EFR encoding, whereas clause 6 describes the decoding procedures. Clause 7 describes variables, constants and tables of the C-code of the GSM EFR codec.

### 4.1 Functional description of audio parts

The analogue-to-digital and digital-to-analogue conversion will in principle comprise the following elements:

1) analogue to uniform digital PCM:

- microphone;
- input level adjustment device;
- input anti-aliasing filter;
- sample-hold device sampling at 8 kHz;
- analogue-to-uniform digital conversion to 13-bit representation.

The uniform format shall be represented in two's complement.

2) uniform digital PCM to analogue:

- conversion from 13-bit/8 kHz uniform PCM to analogue;
- a hold device;
- reconstruction filter including  $x/\sin(x)$  correction;