

## SLOVENSKI STANDARD oSIST ISO/DIS 23032:2021

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#### Meteorologija - Daljinsko zaznavanje vetra na tleh - Radar za profiliranje vetra

Meteorology - Ground-based remote sensing of wind - Radar wind profiler

Météorologie - Télédétection du vent basée au sol - Profileur de vent

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## Meteorology — Ground-based remote sensing of wind — Radar wind profiler

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#### Foreword

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### Introduction

Radar wind profiler, also referred to as wind profiler radar, wind profiling radar, atmospheric radar, or clear-air Doppler radar (hereafter abbreviated to WPR) is an instrument that measures height profiles of wind velocity in clear air. WPR detects echoes produced by perturbations of the radio refractive index with a scale half of the radar wavelength (i.e., Bragg scale). The mechanism of radio wave scattering in clear air was theoretically and experimentally understood in the 1960s. Since the 1970s, large-sized Doppler radars for observing wind and turbulence in the mesosphere, stratosphere, and the troposphere (MST radars) have been developed. Owing to their capability of measuring wind and turbulence with excellent time and height resolution, they have made great contributions to describing and clarifying the dynamical processes in the atmosphere.

Based on the MST radars, WPRs have been developed mainly since the 1980s. WPRs are designed for measuring wind velocity predominantly in the troposphere, including the atmospheric boundary layer. The measurement principle of WPRs are the same used in MST radars but a WPR is frequently smaller in size than a typical MST radar. WPR can measure wind profiles in both a clear and cloudy atmosphere.

In order to monitor and forecast meteorological phenomena, nationwide operational WPR networks have been constructed by meteorological agencies. Operational WPRs contribute to improving weather forecast accuracy through assimilation of their wind products into numerical weather prediction models used by meteorological agencies. Wind products obtained by operational WPRs are distributed globally. Further applications of WPRs include the measurement of wind profiles in the vicinity of airports to enable or improve wind shear warnings. The use of WPRs can improve an airport's ability to safely depart and land aircraft. WPRs are also used to analyse or predict the diffusion of pollutants. In addition, WPRs are widely used by government agencies and various industries, including chemical plants, mines, and power plants, to control emission levels or for computation of nowcast trajectories during emergency situations. The high-quality wind products of WPRs are also widely used in atmospheric research. Therefore, WPRs are an indispensable means for observing wind profiles continuously in time and height. By additionally using radioacoustic sounding system, WPRs can measure height profiles of virtual temperaturelog/standards/sist/623d8bee-3a01-47b0-8bbc-

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In order to attain and retain high quality wind products, WPRs shall be designed, manufactured, and maintained with state-of-the-art knowledge and ensured measurement capability. Aiming at ensuring measurement capability of WPRs, this International Standard provides guidelines in design, manufacture, installation, and maintenance of WPRs.

# Meteorology — Ground-based remote sensing of wind — Radar wind profiler

#### 1 Scope

This International Standard provides guidelines for the design, manufacture, installation, and maintenance of a WPR. It describes the followings.

- Measurement principle (<u>Clause 5</u>). Scatterers that produce echoes and methods of wind velocity measurement are described. The description of the measurement principle mainly aims at providing the information necessary for describing the guidelines in <u>sections 6–11</u>.
- Guidelines for WPR system (<u>Clause 6</u>). Frequency control and stability, hardware, software, and signal processing are described. They are mainly applied in designing and manufacturing the hardware and software of WPR.
- Guidelines for system performance (<u>Clause 7</u>). Measurement resolution, range sampling, radar sensitivity evaluation, and measurement accuracy are described. They can be used for estimating the measurement performance of a WPR's system design and operation.
- Guidelines for quality control (QC: Clause &) RD PREVIEW
- Guidelines for measurement products and data format (<u>Clause 9</u>). Measurement products obtained by a WPR and their data levels are defined. Guidelines for data file formats are also described.
- Guidelines for installation (<u>Clause 10</u>) and <u>maintenance</u> (<u>Clause 11</u>).

https://standards.iteh.ai/catalog/standards/sist/623d8bee-3a01-47b0-8bbc-This international standard does not aim at providing a thorough description of the measurement principle, WPR systems, and WPR applications. For further details of these items, readers should refer to technical books (e.g., [1][2][3]).

WPRs are referred to by various names (e.g., radar wind profiler, wind profiler radar, wind profiling radar, atmospheric radar, or clear-air Doppler radar). Conventional naming for WPRs should be allowed.

#### 2 Normative references

There are no normative references in this document.

#### 3 Terms and definitions

No terms and definitions are listed in this document.

#### 4 Abbreviated terms

#### 4.1 Definition of abbreviations

ACS	adaptive clutter suppression
A/D	analog-to-digital
ADC	A/D converter
BUFR	binary universal form for the representation of meteorological data
СОНО	coherent oscillator

CRI	coherent radar imaging
D/A	digital-to-analog
DBS	Doppler beam swinging
DCMP	directionally constrained minimization of power
DSP	digital signal processor
FCA	Full correlation analysis
FDI	frequency domain interferometry
FMCW	frequency modulated continuous wave
I/0	input / output
IF	intermediate frequency
FPGA	field programmable gate array
IPP	inter pulse period
ITU	International Telecommunication Union
JMA	Japan Meteorological Agency
LNA	low noise amplifier
NC-DCMP	norm-constrained DCMP
NF	noise figure
QC	quality control
RF	radio frequency STANDARD PREVIEW
RIM	range imaging (stondards itch ai)
RL	antenna return loss
SA	spaced antenna osist iso/DIS 23032-2021
SNR	signal to/noise datio ai/catalog/standards/sist/623d8bee-3a01-47b0-8bbc-
STALO	stable (stabilized) 16cal oscillatoriso-dis-23032-2021
UHF	ultra high frequency
UPS	uninterruptible power supply
VHF	very high frequency
VAD	velocity azimuth display
VSWR	voltage standing wave ratio
WMO	World Meteorological Organization
WPR	radar wind profiler, wind profiler radar, wind profiling radar, atmospheric radar, or clear-air Doppler radar

#### 4.2 Definition of symbols

С	speed of light ( $\simeq 3,0 \times 10^8 \text{ m s}^{-1}$ )
$C_{\rm n}^2$	refractive index structure constant
f <sub>Nyq</sub>	Nyquist frequency
<i>f</i> <sub>r</sub>	mean Doppler frequency shift of the echo
G <sub>ant</sub>	antenna gain in decibels
L <sub>p</sub>	loss factor caused by the pulse shaping
n	radio refractive index
N <sub>beam</sub>	number of antenna beam directions

N <sub>coh</sub>	number of coherent integrations. In this document, N <sub>coh</sub> is defined as the number
	excluding N <sub>pseq</sub>
N <sub>data</sub>	number of elements in I/Q time series
N <sub>freq</sub>	number of transmitted frequencies
N <sub>incoh</sub>	number of incoherent integrations
N <sub>pseq</sub>	number of pulse sequences
N <sub>subp</sub>	number of sub-pulses used in phase-modulated pulse compression
T <sub>IPP</sub>	inter pulse period
P <sub>echo</sub>	echo power
P <sub>N</sub>	noise power of the receiver
P <sub>n</sub>	noise power of the Doppler spectrum
p <sub>n</sub>	noise power of the Doppler spectrum per Doppler velocity bin
P <sub>p</sub>	peak output power of the transmitter
Pt	peak output power at the antenna
u	zonal wind velocity
v	meridional wind velocity RD PREVIEW
V <sub>pp</sub>	peak-to-peak voltage
V <sub>r</sub>	radial Doppler velocity
V <sub>s</sub>	sample volume <u>SIST ISO/DIS 23032:2021</u>
V <sub>wind</sub>	wind vector 65d414470/osist-iso-dis-23032-2021
W	vertical wind velocity
Δr	range resolution
η	volume reflectivity
λ	radar wavelength
$ au_{3dB}$	time width between the two 3-dB drop-off points from the peak point
$\tau_{\rm d}$	duration during which the transmission signal is generated
τ <sub>p</sub>	transmitted pulse width
Н	Hermitian operator (complex transposition)
Т	superscript which indicates matrix transposition.
*	complex conjugation

#### 5 Measurement principle

#### 5.1 Spectral parameters of the echo

The properties of all WPR echoes are generally estimated from the properties of the Doppler spectrum<sup>1</sup>). Spectral analysis is typically applied to estimate a finite set of parameters such as signal to noise ratio

<sup>1)</sup> For real-time signal processing to obtain the Doppler spectrum, see sections 6.2.2 and 6.2.6.

(SNR), Doppler shift and spectral (spectrum) width<sup>2</sup>). Of particular importance for a WPR is the echo generated by clear air scattering (clear-air echo). For details of the clear-air echo, see <u>5.2.1</u>.

The frequency distribution of the echo contains information on the radial Doppler velocity ( $V_r$ ,) and on the wind variance caused by turbulence. Figure 1 shows an example of the Doppler spectrum. In general, it is assumed that the Doppler spectrum of the echo ( $S_{echo}$ ) follows a Gaussian distribution<sup>3</sup>).

This assumption is generally applied for the clear-air echo. In this assumption, only the zeroth, first, and second order moments of the echo are taken into account when determining the spectral parameters. In the event of deviations from this assumption, higher order moments may be considered. The noise produced in the receiver (receiver noise) can generally be regarded as white noise. For details of the receiver noise, see <u>6.2.5.4</u>.



#### Кеу

- X doppler velocity
- Y intensity

#### NOTE

- For the definition of the symbols which are not listed in the keys, see text.
- The thin solid curve is an example of a Doppler spectrum which contains the Doppler spectrum of the echo ( $S_{echo}$ ) and the white noise. The thick solid curve is the sum of the noise power of the Doppler spectrum ( $P_n$ ) and the idealized  $S_{echo}$  which follows a Gaussian distribution and does not have perturbation. The idealized  $S_{echo}$  and  $P_n$  are darkly and lightly shaded, respectively. The power of the idealized  $S_{echo}$  is denoted by  $P_{echo}$ .  $B_s$  is the frequency bandwidth of the Doppler spectrum. Mean Doppler frequency shift ( $f_r$ ), spectral width defined as the standard deviation ( $\sigma_{std}$ ), spectral width defined as the half-power full width ( $\sigma_{3dB}$ ), and the peak intensity of the idealized  $S_{echo}$  ( $p_k$ ) are indicated by arrows.
- The received  $S_{echo}$  and the noise shown in this figure were produced by a numerical simulation. In the numerical simulation, Doppler spectra composed of  $S_{echo}$  and white noise were produced. It is assumed that

<sup>2)</sup> Interchangeable with spectral width, spectrum width, is also frequently used. The two terms have the same meaning.

<sup>3)</sup> This assumption shall be carefully discriminated from the assumption that the received signal is the realization of one or more Gaussian stochastic processes, which include those in both radio wave scattering and of course, uncorrelated (white) noise.

each spectrum point of  $S_{echo}$  follows the  $\chi^2$  distribution with 2 degrees of freedom. Produced Doppler spectra were integrated, and the Doppler spectrum after the integration (i.e., incoherent integration) is plotted. Therefore, the noise variance over  $B_s$  is smaller than the square of the noise power per Doppler velocity bin  $(p_n^2)$ .

The noise variance is one of the principal factors that determine the sensitivity of a WPR receiver. See sections 6.2.2 and 7.3 for details of incoherent integration and radar sensitivity, respectively.

#### Figure 1 — Example of Doppler spectrum and spectral parameters.

Echo power ( $P_{echo}$ ),  $V_r$ , and the spectral width are the principal parameters that characterizes the echo. They are referred to as the spectral parameters.  $V_r$  is computed from the mean Doppler frequency shift of the echo ( $f_r$ ).  $P_{echo}$  and  $f_r$  are also the zeroth and first order moment of  $S_{echo}$ , respectively. The spectral width defined as the standard deviation ( $\sigma_{std}$ ) is the square root of the second order moment of  $S_{echo}$  (see Figure 1).  $P_{echo}$ ,  $f_r$ , and  $\sigma_{std}$  are expressed by

$$P_{\rm echo} = \int S_{\rm echo} \left( f \right) df \,, \tag{1}$$

$$f_{\rm r} = \frac{\int fS_{\rm echo}(f)df}{\int S_{\rm echo}(f)df}, \text{ and}$$
(2)

$$\sigma_{\rm std} = \sqrt{\frac{\int (f - f_{\rm r})^2 S_{\rm echo}(f) df}{\int S_{\rm echo}(f) df}} ({\rm standards.iteh.ai})$$
(3)

where *f* is the Doppler frequency. <u>oSIST ISO/DIS 23032:2021</u> https://standards.iteh.ai/catalog/standards/sist/623d8bee-3a01-47b0-8bbc-The relation between  $f_r$  and  $V_r$ 3is expressed by iso-dis-23032-2021

$$V_{\rm r} \approx -\frac{\lambda}{2} f_{\rm r} \,. \tag{4}$$

In Equation (4),  $V_r$  is defined to be positive when its direction is away from the antenna. However, when one prefers to use the sign definition of  $V_r$  as that of  $f_r$ ,  $V_r$  can be defined to be positive when its direction is toward the antenna. In any case, the direction of  $V_r$  shall be defined clearly in the design and manufacture of the WPR in order to prevent possible mistakes in the design, manufacture, operation, and maintenance of the WPR. In this document,  $V_r$  is defined to be positive when its direction is away from the antenna.

The spectral width can be also defined as the half-power full width ( $\sigma_{
m 3dB}$ ) or half-power half width of

the echo (i.e.,  $\frac{\sigma_{3dB}}{2}$ ). When  $S_{echo}$  is assumed to follow a Gaussian distribution,  $\sigma_{3dB}$  can be calculated by the relation

$$\sigma_{\rm 3dB} = 2\sqrt{2\ln 2} \,\,\sigma_{\rm std} \tag{5}$$

Because the spectral width can be expressed under the above-mentioned definitions, the definition of the spectral width shall be given explicitly. It shall be noted that the spectral width is not only determined by wind perturbation caused by turbulence, but also contains broadening effects due to the angular and vertical extension of the sample volume [4]. Details of the sample volume are described in 7.1.2.

In the estimation of the spectral parameters, the noise power of the Doppler spectrum ( $P_n$ ) is also estimated. SNR is expressed by

$$SNR = \frac{P_{echo}}{P_{n}}$$
(6)

In the digital signal processing for estimating the spectral parameters and  $P_n$ , the noise power per Doppler velocity bin  $(p_n)$  is generally used.  $p_n$  is expressed by

$$p_{\rm n} = P_{\rm n} \, \frac{\Delta f}{B_{\rm s}},\tag{7}$$

where

is the frequency bandwidth of the Doppler spectrum;  $B_{c}$ 

 $\Delta f$ is the frequency resolution of the Doppler spectrum (i.e., interval of the Doppler frequency bins).

It is noted that interference from other radio sources that contaminates the received signal has frequency dependency in general. Therefore, contamination due to the radio interference can produce a frequency dependency of the noise. Details of the interference from radio sources are described in sections 5.2.4 and 10.6.

When it is assumed that  $S_{echo}$  follows a Gaussian distribution and SNR is infinite, the estimation error of Doppler velocity or the spectral width,  $\varepsilon_v$ , can be estimated by

$$\varepsilon_{v} = K_{v} \left(\frac{\sigma_{v}}{T_{c}}\right)^{\frac{1}{2}},$$
ere
$$\frac{(\text{Standards.iten.al})}{\frac{\text{oSIST ISO/DIS 23032:2021}}{\frac{\text{oSIST ISO/DIS 23032:2021}}{\frac{1}{1}}}$$

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(8)

where

 $K_{\rm w}$ is the coefficient;

 $\sigma_{\rm v}$ is the spectral width defined as the standard deviation in  $m s^{-1}$ ;

is the measurement period in s.  $T_{c}$ 

When the antenna beam direction is changed after collecting a Doppler spectrum (i.e., after  $N_{pseq}N_{coh}N_{data}$  times transmissions and receptions) or after collecting all of the Doppler spectra used in incoherent integration (i.e., after  $N_{pseq}N_{coh}N_{data}N_{incoh}$  times transmissions and receptions),  $T_{c} = T_{IPP} N_{pseq} N_{coh} N_{data} N_{incoh}$ . When the antenna beam direction is changed on a pulse-to-pulse basis,  $T_{\rm c} = T_{\rm IPP} N_{\rm beam} N_{\rm pseq} N_{\rm coh} N_{\rm data} N_{\rm incoh}$ . See <u>6.2.3.2.5</u> for details about the timing change of the antenna beam direction.

 $K_{\rm w}$  is defined by

$$K_{\rm v} = k_{\rm err} \left(\frac{\lambda}{2}\right)^{\frac{1}{2}},\tag{9}$$

where

is the coefficient; k<sub>err</sub>

λ is the radar wavelength; )

Equations (8) and (9) are referred from Equations (13) and (14) of  $k_{arr}$ from <u>Table 1</u> of [5] is listed in <u>Table 1</u>.

Parameter	Least square method	Moment method
Doppler velocity	0,63	0,38
Spectral width	0,60	0,24

Fable 1 — Value of $k_{a}$
----------------------------

Error estimations of the spectral parameters when considering SNR is described in sections 6.3, 6.4, and 6.5 of<sup>[2]</sup>.

#### 5.2 Sources of received signals

#### 5.2.1 **Turbulent scattering and partial reflection**

The ability to detect the clear-air echo<sup>4</sup>) is the most important characteristic of a WPR. It makes a WPR capable of determining vertically resolved profiles of the wind vector from the measured Doppler shift of the clear-air echo. There are two major mechanisms that produce echoes in clear air: turbulent scattering from atmospheric turbulence and partial reflection from the horizontally stratified atmosphere. Partial reflection is also referred to as Fresnel scattering. Atmospheric turbulence produces perturbation of n, and perturbations of n with the scale of half of \* (i.e., Bragg scale) is a source of radio wave scattering in clear air

*n* in the neutral (i.e., unionized) atmosphere is given by  

$$n = 1+7,76 \times 10^{-5} \frac{p}{T} + 3,73 \times 10^{-1} \frac{e}{T_{c}^{2} \text{ SIST ISO/DIS 23032:2021}}$$
(10)  
https://standards.iteh.ai/catalog/standards/sist/623d8bee-3a01-47b0-8bbc-  
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T.

- р is the atmosphere pressure in hPa;
- Т is the atmospheric temperature in K;
- е is the partial pressure of water vapour.

When perturbations of *n* is isotropic, turbulent scattering is also isotropic.

The refractive index structure constant  $C_n^2$  is defined as

$$\overline{[n(r+\delta r)-n(r)]^2} = C_n^2 |\delta r|^{2/3},$$
(11)

where r is an arbitrary position and  $\delta r$  is a small distance between two spaced locations, respectively. Because T and e are perturbed by turbulence and n depends on them,  $C_n^2$  significantly varies due to the atmospheric conditions that determine T and e (see Equation (10)).

The frequency of a WPR is generally selected so that turbulent scattering occurs in the inertial subrange of turbulence. Frequencies between 50 MHz and 3 GHz have generally been used for WPRs.

In the inertial sub-range, the energy cascades from the largest eddies to the smallest ones through an inertial (and inviscid) mechanism. The inertial sub-range exists between the inner scale of turbulence (  $l_0$  ) and the buoyancy length scale (  $L_B$  ).  $l_0$  is the scale for determining the transition region between the viscous and inertial sub-ranges, and  $L_{R}$  is the scale for determining the transition region between

<sup>4)</sup> The clear-air echo is a return from a radio wave scattering caused by variations of the radio refractive index *n*, and does not include scatterings from hard targets in the air (e.g., hydrometeors, insects, birds, and aircrafts).