SIST

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Meteorologija - Daljinsko zaznavanje vetra na tleh - Radar za profiliranje vetra

Meteorology - Ground-based remote sensing of wind - Radar wind profiler

Météorologie - Télédétection du vent basée au sol - Profileur de vent

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INTERNATIONAL STANDARD

ISO 23032

Meteorology — Ground-based remote sensing of wind — Radar wind profiler

Météorologie — Télédétection du vent basée au sol — Profileur de vent radar

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Contents

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Foreword

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The procedures used to develop this document and those intended for its further maintenance are described in the ISO/IEC Directives, Part 1. In particular, the different approval criteria needed for the different types of ISO documents should be noted. This document was drafted in accordance with the editorial rules of the ISO/IEC Directives, Part 2 (see www.iso.org/directives).

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This document was prepared by Technical Committee ISO/TC 146, *Air quality*, Subcommittee SC 5, *Meteorology*, and by the World Meteorological Organization (WMO) as a common ISO/WMO Standard under the Agreement on Working Arrangements signed between the WMO and ISO in 2008.

Any feedback or questions on this document should be directed to the user's national standards body. A complete listing of these bodies can be found at www.iso.org/members.html.

ISO 23032:2022(E)

Introduction

Radar wind profiler, also referred to as wind profiler radar, wind profiling radar, atmospheric radar, or clear-air Doppler radar (hereafter abbreviated to WPR) is an instrument that measures height profiles of wind velocity in clear air. WPR detects echoes produced by perturbations of the radio refractive index with a scale half of the radar wavelength (i.e. Bragg scale). The mechanism of radio wave scattering in clear air was theoretically and experimentally understood in the 1960s. Since the 1970s, large-sized Doppler radars for observing wind and turbulence in the mesosphere, stratosphere, and the troposphere (MST radars) have been developed. Owing to their capability of measuring wind and turbulence with excellent time and height resolution, they have made great contributions to describing and clarifying the dynamical processes in the atmosphere.

Based on the MST radars, WPRs have been developed mainly since the 1980s. WPRs are designed for measuring wind velocity predominantly in the troposphere, including the atmospheric boundary layer. The measurement principle of WPRs are the same used in MST radars but a WPR is frequently smaller in size than a typical MST radar. WPR can measure wind profiles in both a clear and cloudy atmosphere.

In order to monitor and forecast meteorological phenomena, nationwide operational WPR networks have been constructed by meteorological agencies. Operational WPRs contribute to improving weather forecast accuracy through assimilation of their wind products into numerical weather prediction models used by meteorological agencies. Wind products obtained by operational WPRs are distributed globally. Further applications of WPRs include the measurement of wind profiles in the vicinity of airports to enable or improve wind shear warnings. The use of WPRs can improve an airport's ability to safely depart and land aircraft. WPRs are also used to analyse or predict the diffusion of pollutants. In addition, WPRs are widely used by government agencies and various industries, including chemical plants, mines, and power plants, to control emission levels or for computation of nowcast trajectories during emergency situations. The high-quality wind products of WPRs are also widely used in atmospheric research. Therefore, WPRs are an indispensable means for observing wind profiles continuously in time and height. By additionally using radio acoustic sounding system, WPRs can measure height profiles of virtual temperature.

In order to attain and retain high quality wind products, WPRs need to be designed, manufactured, and maintained with state-of-the-art knowledge and ensured measurement capability. Aiming at ensuring measurement capability of WPRs, this document provides guidelines in design, manufacture, installation, and maintenance of WPRs.

INTERNATIONAL STANDARD   ISO 23032:2022(E)

Meteorology — Ground-based remote sensing of wind — Radar wind profiler

1   Scope

This document provides guidelines for the design, manufacture, installation, and maintenance of a WPR. It describes the following:

— Measurement principle (Clause 5). Scatterers that produce echoes and methods of wind velocity measurement are described. The description of the measurement principle mainly aims at providing the information necessary for describing the guidelines in Clauses 6 to 11.

— Guidelines for WPR system (Clause 6). Frequency, hardware, software, and signal processing are described. They are mainly applied in designing and manufacturing the hardware and software of WPR.

— Guidelines for system performance (Clause 7). Measurement resolution, range sampling, radar sensitivity evaluation, and measurement accuracy are described. They can be used for estimating the measurement performance of a WPR's system design and operation.

— Guidelines for quality control (QC) in digital signal processing (Clause 8).

— Guidelines for measurement products and data format (Clause 9). Measurement products obtained by a WPR and their data levels are defined. Guidelines for data file formats are also described.

— Guidelines for installation (Clause 10) and maintenance (Clause 11).

This document does not aim at providing a thorough description of the measurement principle, WPR systems, and WPR applications. For further details of these items, users are referred to technical books (e.g. References [1],[2],[3]).

WPRs are referred to by various names (e.g. radar wind profiler, wind profiler radar, wind profiling radar, atmospheric radar, or clear-air Doppler radar). Conventional naming for WPRs should be allowed.

2   Normative references

There are no normative references in this document.

3   Terms and definitions

No terms and definitions are listed in this document.

ISO and IEC maintain terminology databases for use in standardization at the following addresses:

— ISO Online browsing platform: available at https://www.iso.org/obp

— IEC Electropedia: available at https://www.electropedia.org/

ISO 23032:2022(E)

4 Symbols and abbreviated terms

4.1 Symbols

c	speed of light ($\approx 3{,}0 \times 10^8$ m s^{-1})
C_n^2	refractive index structure constant
f_{Nyq}	Nyquist frequency
f_r	mean Doppler frequency shift of the echo
G_{ant}	antenna gain in decibels
L_p	loss factor caused by the pulse shaping
n	radio refractive index
N_{beam}	number of antenna beam directions
N_{coh}	number of coherent integrations. In this document, N_{coh} is defined as the number excluding N_{pseq}
N_{data}	number of elements in I and Q (I/Q) time series after coherent integrations. N_{data} is also the number of elements in the Doppler spectrum
N_{freq}	number of transmitted frequencies
N_{incoh}	number of incoherent integrations
N_{pseq}	number of pulse sequences
N_{subp}	number of sub-pulses used in phase-modulated pulse compression
T_{IPP}	inter pulse period
P_{echo}	echo power
P_N	noise power of the receiver
P_n	noise power of the Doppler spectrum
p_n	noise power of the Doppler spectrum per Doppler velocity bin
P_p	peak output power of the transmitter
P_t	peak output power at the antenna
u	zonal wind velocity
v	meridional wind velocity
V_{pp}	peak-to-peak voltage
V_r	radial Doppler velocity
V_s	sample volume
$\boldsymbol{V_{wind}}$	wind vector
w	vertical wind velocity
Δr	range resolution
η	volume reflectivity
λ	radar wavelength
σ_{3dB}	spectral width defined as the half-power full width
σ_{std}	spectral width defined as the standard deviation

$\tau_{3\text{dB}}$	time width between the two 3-dB drop-off points from the peak point
τ_{d}	duration during which the transmission signal is generated
τ_{p}	transmitted pulse width
H	Hermitian operator (complex transposition)
T	superscript which indicates matrix transposition
*	complex conjugation

4.2 Abbreviated terms

ACS	adaptive clutter suppression
A/D	analog-to-digital
ADC	A/D converter
BUFR	binary universal form for the representation of meteorological data
COHO	coherent oscillator
CRI	coherent radar imaging
D/A	digital-to-analog
DBS	Doppler beam swinging
DCMP	directionally constrained minimization of power
DSP	digital signal processor
FCA	full correlation analysis
FDI	frequency domain interferometry
FMCW	frequency modulated continuous wave
I/O	input/output
I/Q	in-phase (I)/quadrature-phase (Q)
IF	intermediate frequency
FPGA	field programmable gate array
IPP	inter pulse period
ITU	International Telecommunication Union
JMA	Japan Meteorological Agency
LNA	low noise amplifier
MTBF	mean time between failures
MTTF	mean time to failure
NC-DCMP	norm-constrained DCMP
NF	noise figure
QC	quality control
RF	radio frequency
RIM	range imaging
RL	antenna return loss
SA	spaced antenna
SNR	signal to noise ratio
STALO	stable (stabilized) local oscillator
UHF	ultra high frequency
UPS	uninterruptible power supply

ISO 23032:2022(E)

VHF	very high frequency
VAD	velocity azimuth display
VSWR	voltage standing wave ratio
WMO	World Meteorological Organization
WPR	radar wind profiler, wind profiler radar, wind profiling radar, atmospheric radar, or clear-air Doppler radar

5 Measurement principle

## 5.1 Spectral parameters of the echo

The properties of all WPR echoes are generally estimated from the properties of the Doppler spectrum. Spectral analysis is typically applied to estimate a finite set of parameters such as signal to noise ratio (SNR), Doppler shift and spectral (spectrum) width. Of particular importance for a WPR is the echo generated by clear air scattering (clear-air echo). For details of the clear-air echo, see 5.2.1.

NOTE 1   For real-time signal processing to obtain the Doppler spectrum, see 6.2.2 and 6.2.6.

NOTE 2   Interchangeable with spectral width, spectrum width, is also frequently used. The two terms have the same meaning.

The frequency distribution of the echo contains information on the radial Doppler velocity (V_r) and on the wind variance caused by turbulence. Figure 1 shows an example of the Doppler spectrum. The Doppler spectrum of the echo (S_{echo}) and the noise shown in Figure 1 were produced by a numerical simulation. In the numerical simulation, Doppler spectra composed of S_{echo} and white noise were produced. The noise power of the Doppler spectrum is expressed by P_n. It is assumed that S_{echo} follows a Gaussian distribution and that each spectrum point of S_{echo} follows the χ^2 distribution with 2 degrees of freedom. The frequency bandwidth of the Doppler spectrum is expressed by B_s. Produced Doppler spectra were integrated, and the Doppler spectrum after the integration (i.e. incoherent integration) is plotted. Therefore, the noise variance over B_s is smaller than the square of the noise power per Doppler velocity bin (p_n^2). The noise variance is one of the principal factors that determine the sensitivity of a WPR receiver. See 6.2.2 and 7.3 for details of incoherent integration and radar sensitivity, respectively.

In general, it is assumed that S_{echo} follows a Gaussian distribution. This assumption is generally applied for the clear-air echo. In this assumption, only the zeroth, first, and second order moments of the echo are taken into account when determining the spectral parameters. This assumption shall be carefully discriminated from the assumption that the received signal is the realization of one or more Gaussian stochastic processes, which include those in both radio wave scattering and of course, uncorrelated (white) noise. In the event of deviations from this assumption, higher order moments may be considered. The noise produced in the receiver (receiver noise) can generally be regarded as white noise. For details of the receiver noise, see 6.2.5.4.
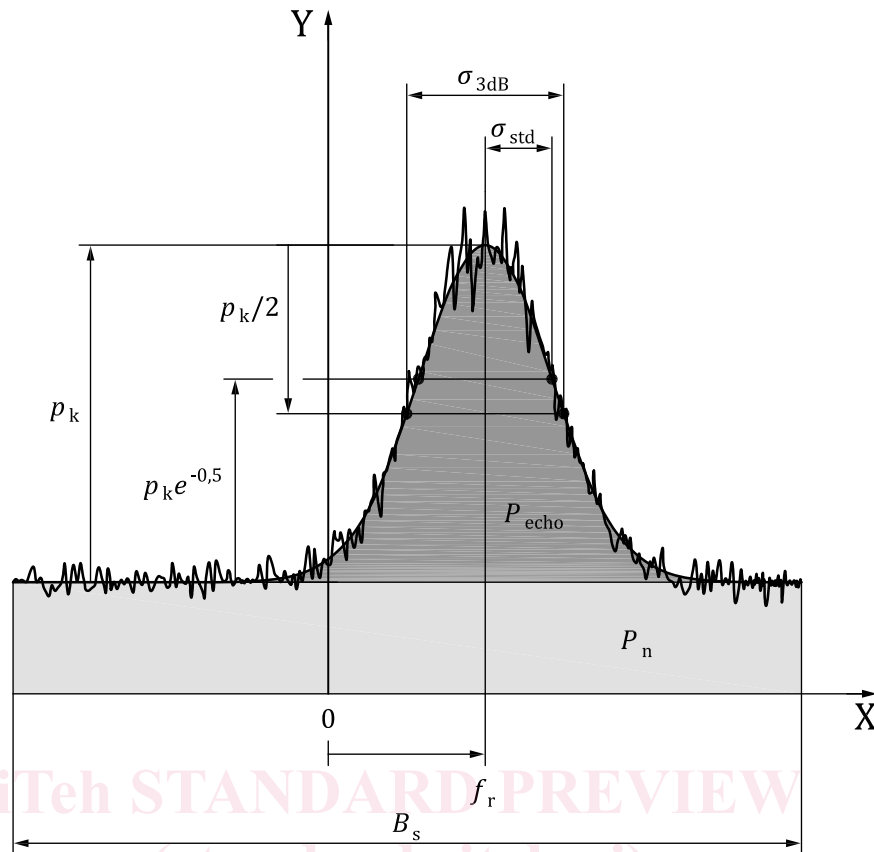
Key

X   Doppler velocity

Y   intensity

NOTE

—   For the definition of the symbols which are not listed in the keys, see text.

—   The thin solid curve is an example of a Doppler spectrum which contains the Doppler spectrum of S_{echo} and the white noise. The thick solid curve is the sum of P_n and the idealized S_{echo} which follows a Gaussian distribution and does not have perturbation. The idealized S_{echo} and P_n are darkly and lightly shaded, respectively. The power of the idealized S_{echo} is denoted by P_{echo}. f_r, σ_{std}, σ_{3dB}, and the peak intensity of the idealized S_{echo} (p_k) is indicated by arrows.

Figure 1 — Example of Doppler spectrum and spectral parameters

Echo power (P_{echo}), V_r, and the spectral width are the principal parameters that characterizes the echo. They are referred to as the spectral parameters. V_r is computed from the mean Doppler frequency shift of the echo (f_r). P_{echo} and f_r are also the zeroth and first order moment of S_{echo}, respectively. The spectral width defined as the standard deviation (σ_{std}) is the square root of the second order moment of S_{echo} (see Figure 1). P_{echo}, f_r, and σ_{std} are expressed by Formula (1), (2) and (3):

$$P_{echo} = \int S_{echo}(f)\,df \tag{1}$$

$$f_r = \frac{\int f S_{echo}(f)\,df}{\int S_{echo}(f)\,df} \tag{2}$$

ISO 23032:2022(E)

$$\sigma_{std} = \sqrt{\frac{\int (f - f_r)^2 S_{echo}(f) df}{\int S_{echo}(f) df}} \tag{3}$$

where f is the Doppler frequency.

The relation between f_r and V_r is expressed by Formula (4):

$$V_r \approx -\frac{\lambda}{2} f_r \tag{4}$$

In Formula (4), V_r is defined to be positive when its direction is away from the antenna. However, when one prefers to use the sign definition of V_r as that of f_r, V_r can be defined to be positive when its direction is toward the antenna. In any case, the direction of V_r shall be defined clearly in the design and manufacture of the WPR in order to prevent possible mistakes in the design, manufacture, operation, and maintenance of the WPR. In this document, V_r is defined to be positive when its direction is away from the antenna.

The spectral width can be also defined as the half-power full width (σ_{3dB}) or half-power half width of

the echo (i.e. $\frac{\sigma_{3dB}}{2}$). When S_{echo} is assumed to follow a Gaussian distribution, σ_{3dB} can be calculated

by the relation of Formula (5):

$$\sigma_{3dB} = 2\sqrt{2\ln 2} \; \sigma_{std} \tag{5}$$

Because the spectral width can be expressed under the above-mentioned definitions, the definition of the spectral width shall be given explicitly. It shall be noted that the spectral width is not only determined by wind perturbation caused by turbulence, but also contains broadening effects due to the angular and vertical extension of the sample volume[4]. Details of the sample volume are described in 7.1.2.

In the estimation of the spectral parameters, P_n is also estimated. SNR is expressed by Formula (6):

$$SNR = \frac{P_{echo}}{P_n} \tag{6}$$

In the digital signal processing for estimating the spectral parameters and P_n, the noise power per Doppler velocity bin (p_n) is generally used. p_n is expressed by Formula (7):

$$p_n = P_n \frac{\Delta f}{B_s} \tag{7}$$

where Δf is the frequency resolution of the Doppler spectrum (i.e. interval of the Doppler frequency bins). It is noted that interference from other radio sources that contaminates the received signal has frequency dependency in general. Therefore, contamination due to the radio interference can produce a frequency dependency of the noise. Details of the interference from radio sources are described in 5.2.4 and 10.6.

When it is assumed that S_{echo} follows a Gaussian distribution and SNR is infinite, the estimation error of Doppler velocity or the spectral width, ε_v, can be estimated by Formula (8):

$$\varepsilon_v = K_v \left(\frac{\sigma_v}{T_c}\right)^{\frac{1}{2}} \tag{8}$$

where

K_v is the coefficient;

σ_v is the spectral width defined as the standard deviation in m s^{-1};

T_c is the measurement period in s.

When the antenna beam direction is changed after collecting a Doppler spectrum (i.e. after $N_{\mathrm{pseq}}N_{\mathrm{coh}}N_{\mathrm{data}}$ times transmissions and receptions) or after collecting all of the Doppler spectra used in incoherent integration (i.e. after $N_{\mathrm{pseq}}N_{\mathrm{coh}}N_{\mathrm{data}}N_{\mathrm{incoh}}$ times transmissions and receptions), $T_c = T_{\mathrm{IPP}}N_{\mathrm{pseq}}N_{\mathrm{coh}}N_{\mathrm{data}}N_{\mathrm{incoh}}$. When the antenna beam direction is changed on a pulse-to-pulse basis, $T_c = T_{\mathrm{IPP}}N_{\mathrm{beam}}N_{\mathrm{pseq}}N_{\mathrm{coh}}N_{\mathrm{data}}N_{\mathrm{incoh}}$. See 6.2.3.2.5 for details about the timing change of the antenna beam direction.

K_v is defined by Formula (9):

$$K_v = k_{\mathrm{err}}\left(\frac{\lambda}{2}\right)^{\frac{1}{2}} \tag{9}$$

where

k_{err} is the coefficient;

λ is the radar wavelength;

Formulae (8) and (9) are derived from Formulae (13) and (14) in Reference [5], respectively. The value of k_{err} (see Reference [5]) is listed in Table 1.

Table 1 — Value of k_{err}

Parameter	Least square method	Moment method
Doppler velocity	0,63	0,38
Spectral width	0,60	0,24

Error estimations of the spectral parameters when considering SNR is described in 6.3, 6.4, and 6.5 of Reference [2].

5.2   Sources of received signals

5.2.1   Turbulent scattering and partial reflection

The ability to detect the clear-air echo is the most important characteristic of a WPR. It makes a WPR capable of determining vertically resolved profiles of the wind vector from the measured Doppler shift of the clear-air echo. There are two major mechanisms that produce echoes in clear air: turbulent scattering from atmospheric turbulence and partial reflection from the horizontally stratified atmosphere. Partial reflection is also referred to as Fresnel scattering. Atmospheric turbulence produces perturbation of n, and perturbations of n with the scale of half of λ (i.e. Bragg scale) is a source of radio wave scattering in clear air.

NOTE   The clear-air echo is a return from a radio wave scattering caused by variations of the radio refractive index n, and does not include scatterings from hard targets in the air (e.g. hydrometeors, insects, birds, and aircrafts).