SLOVENSKI STANDARD
SIST EN 50067:1999

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Specification of the radio data system (RDS) for VHF/FM sound broadcasting in the freqzuency range from 87,5 to 108,0 MHz

Specification of the radio data system (RDS) for VHF/FM sound broadcasting in the frequency range from 87,5 to 108,0 MHz

Spezifikation des Radio-Daten-Systems (RDS) für den VHF/FM Tonrundfunk im Frequenzbereich 87,5 bis 108,0 MHz

Spécifications du système de radiodiffusion de données (RDS) pour la radio à modulation de fréquence dans la bande de 87,5 à 108,0 MHz

Ta slovenski standard je istoveten z: EN 50067:1998

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EN 50067

Specification of the radio data system (RDS)

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This European Standard exists in three official versions (English, French, German). A version in any other language made by translation under the responsibility of a CENELEC member into its own language and notified to the Central Secretariat has the same status as the official versions.

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CENELEC

European Committee for Electrotechnical Standardization
Comité Européen de Normalisation Electrotechnique
Europäisches Komitee für Elektrotechnische Normung

Central Secretariat: rue de Stassart 35, B - 1050 Brussels

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FOREWORD

The Radio Data System RDS was developed by the European Broadcasting Union (EBU) Member countries who collaborated towards an internationally agreed standard for such a system. The Specification of the RDS System was initially published by the EBU in 1984 as doc.Tech 3244 [8] and is also the subject of ITU-R Recommendation 643-2.

This revised text, which is published by the European Committee for Electrotechnical Standardization (CENELEC), was prepared by the RDS Forum in close collaboration with the Technical Committee 207 of CENELEC, and in close collaboration with experts from the EBU. In addition, certain elements of text have been revised to accord with experience gained with the RDS System and changes in broadcasting practice since the Specification was published. It is, nevertheless, expected that receivers produced to accord with this Specification will be compatible with RDS broadcasts which conform with previous editions of this Specification.

Attention is drawn to the fact that there may be Intellectual Property Rights (IPR) in relation to certain provisions of this standard. The technical experts of TC 207 were unable to fully identify such claims due to the complicated legal issues involved. IPR holders should notify CENELEC of their claims.

This document was submitted to the Unique Acceptance Procedure and was approved by CENELEC as EN 50067 on 1998-04-01.

The following dates were fixed:

- latest date by which the EN has to be implemented at national level by publication of an identical national standard or by endorsement	(dop)	1998-12-01
- latest date by which the national standards conflicting with the EN have to be withdrawn	(dow)	1998-12-01

This European Standard replaces EN 50067:1992.

This version of the specification includes several significant enhancements to the RDS features: Open Data Applications, Programme Type Name, EWS and Enhanced Paging Protocol. These are a fully backwards compatible set of additions. A receiver implemented in accordance with EN 50067: 1992 but receiving a transmission in accordance with this standard, whilst not able to respond to the enhancements, will not significantly under perform.

This standard is also drafted to facilitate a world-wide standard by working towards harmonisation with the US NRSC RBDS standard.

CONTENTS

ANNEXES

0 Scope

The Radio Data System, RDS, is intended for application to VHF/FM sound broadcasts in the range 87.5 MHz to 108.0 MHz which may carry either stereophonic (pilot-tone system) or monophonic programmes. The main objectives of RDS are to enable improved funtionality for FM receivers and to make them more user-friendly by using features such as Programme Identification, Programme Service name display and where applicable, automatic tuning for portable and car radios, in particular. The relevant basic tuning and switching information shall therefore be implemented by the type 0 group (see 3.1.5.1), and it is not optional unlike many of the other possible features in RDS.

1 Modulation characteristics of the data channel (physical layer)

The Radio Data System is intended for application to VHF/FM sound broadcasting transmitters in the range 87.5 to 108.0 MHz, which carry stereophonic (pilot-tone system) or monophonic sound broadcasts (see ITU-R Recommendation BS.450-2).

It is important that radio-data receivers are not affected by signals in the multiplex spectrum outside the data channel.

The system can be used simultaneously with the ARI system (see annex H), even when both systems are broadcast from the same transmitter. However, certain constraints on the phase and injection levels of the radio-data and ARI signals must be observed in this case (see 1.2 and 1.3).

The data signals are carried on a subcarrier which is added to the stereo multiplex signal (or monophonic signal as appropriate) at the input to the VHF/FM transmitter. Block diagrams of the data source equipment at the transmitter and a typical receiver arrangement are shown in figures 1 and 2, respectively.

1.1 Subcarrier frequency

During stereo broadcasts the subcarrier frequency will be locked to the third harmonic of the 19-kHz pilot-tone. Since the tolerance on the frequency of the 19-kHz pilot-tone is ± 2 Hz (see ITU-R Recommendation BS.450-2), the tolerance on the frequency of the subcarrier during stereo broadcasts is ± 6 Hz.

During monophonic broadcasts the frequency of the subcarrier will be 57 kHz ± 6 Hz.

1.2 Subcarrier phase

During stereo broadcasts the subcarrier will be locked either in phase or in quadrature to the third harmonic of the 19 kHz pilot-tone. The tolerance on this phase angle is ± 10°, measured at the modulation input to the FM transmitter.

In the case when ARI and radio-data signals are transmitted simultaneously, the phase angle between the two subcarriers shall be 90° ± 10°.

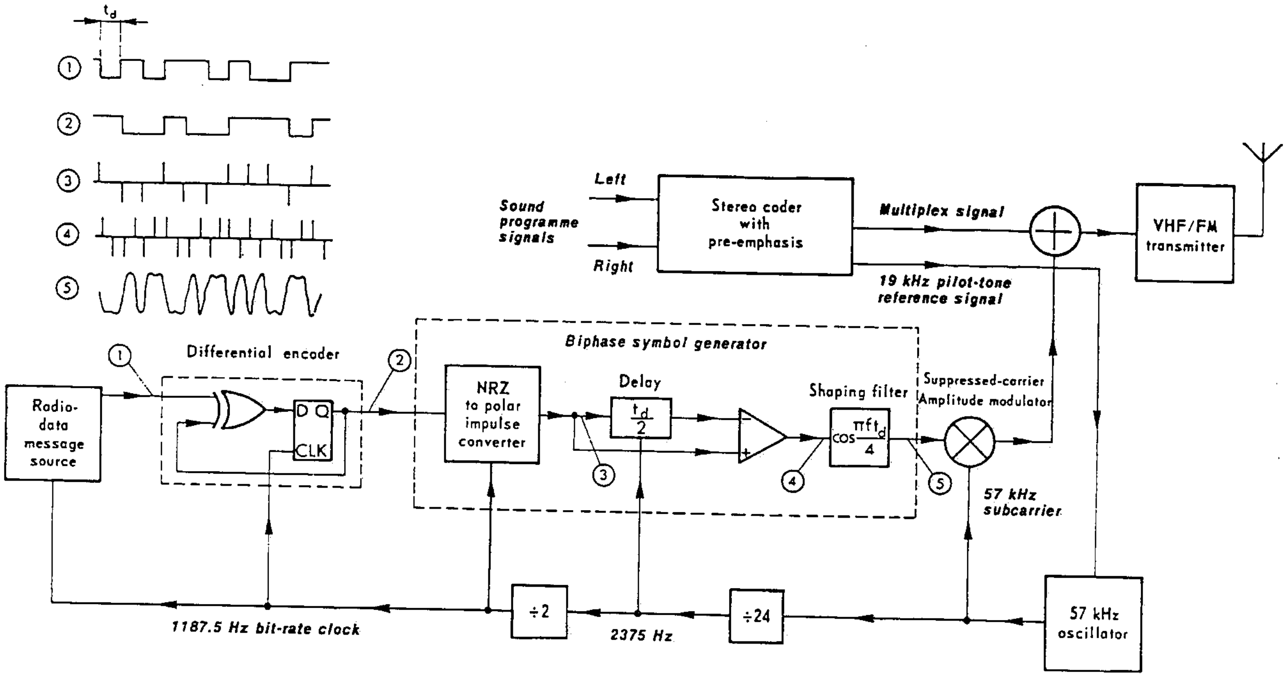
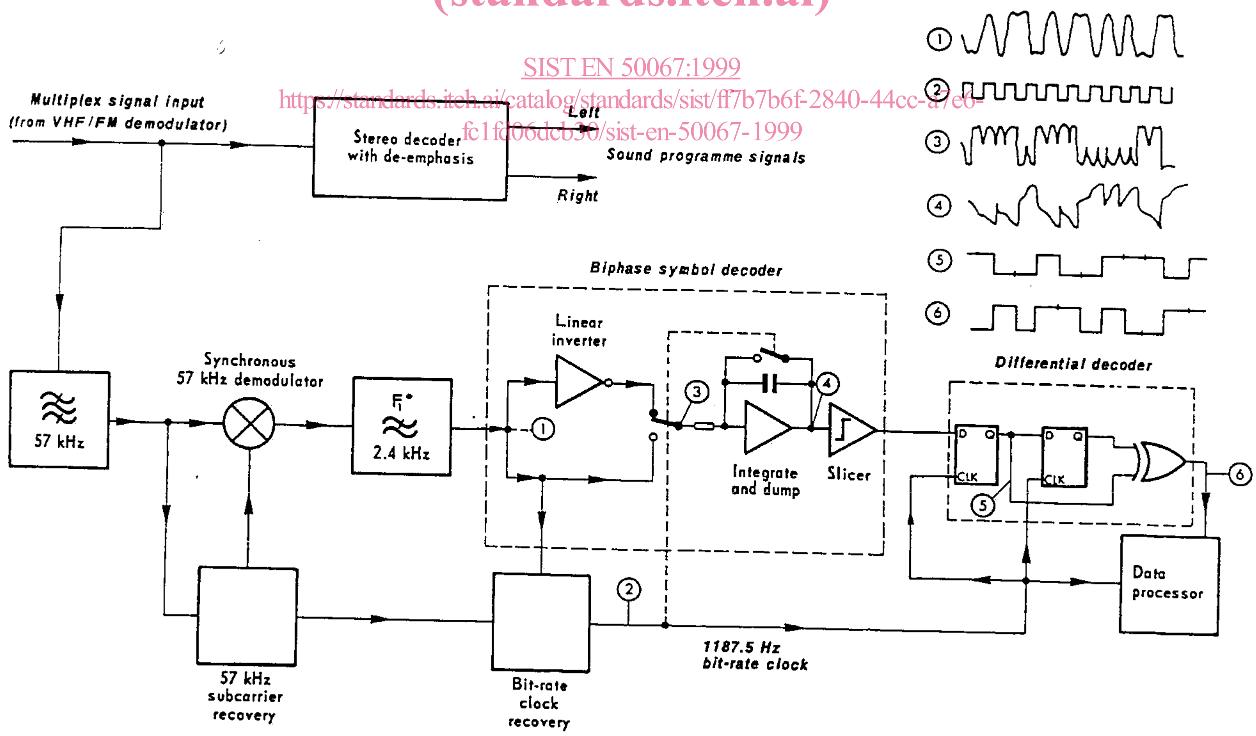


Figure 1: Block diagram of radio-data equipment at the transmitter



The overall data-shaping in this decoder comprises the filter F_1 and the data-shaping inherent in the biphase symbol decoder. The amplitude/frequency characteristic of filter F_1 is, therefore, not the same as that given in figure 3.

Figure 2: Block diagram of a typical radio-data receiver/decoder

1.3 Subcarrier level

The deviation range of the FM carrier due to the unmodulated subcarrier is from ± 1.0 kHz to ± 7.5 kHz. The recommended best compromise is ± 2.0 kHz [1]). The decoder/demodulator should also operate properly when the deviation of the subcarrier is varied within these limits during periods not less than 10 ms.

In the case when ARI (see annex H) and radio-data signals are transmitted simultaneously, the recommended maximum deviation due to the radio-data subcarrier is ± 1.2 kHz and that due to the unmodulated ARI subcarrier should be reduced to ± 3.5 kHz.

The maximum permitted deviation due to the composite multiplex signal is ± 75 kHz.

1.4 Method of modulation

The subcarrier is amplitude-modulated by the shaped and biphase coded data signal (see 1.7). The subcarrier is suppressed. This method of modulation may alternatively be thought of as a form of two-phase phase-shift-keying (psk) with a phase deviation of ± 90°.

1.5 Clock-frequency and data-rate

The basic clock frequency is obtained by dividing the transmitted subcarrier frequency by 48. Consequently, the basic data-rate of the system (see figure 1) is 1187.5 bit/s ± 0.125 bit/s.

1.6 Differential coding

The source data at the transmitter are differentially encoded according to the following rules:

Table 1: Encoding rules

Previous output (at time t_{i-1})	New input (at time t_i)	New output (at time t_i)
0	0	0
0	1	1
1	0	1
1	1	0

where t_i is some arbitrary time and t_{i-1} is the time one message-data clock-period earlier, and where the message-data clock-rate is equal to 1187.5 Hz.

[1]) With this level of subcarrier, the level of each sideband of the subcarrier corresponds to half the nominal peak deviation level of ± 2.0 kHz for an "all-zeroes" message data stream (i.e. a continuous bit-rate sine-wave after biphase encoding).

Thus, when the input-data level is 0, the output remains unchanged from the previous output bit and when an input 1 occurs, the new output bit is the complement of the previous output bit.

In the receiver, the data may be decoded by the inverse process:

Table 2: Decoding rules

Previous input (at time t_{i-1})	New input (at time t_i)	New output (at time t_i)
0	0	0
0	1	1
1	0	1
1	1	0

The data is thus correctly decoded whether or not the demodulated data signal is inverted.

1.7 Data-channel spectrum shaping

The power of the data signal at and close to the 57 kHz subcarrier is minimized by coding each source data bit as a biphase symbol.

This is done to avoid data-modulated cross-talk in phase-locked-loop stereo decoders, and to achieve compatibility with the ARI system. The principle of the process of generation of the shaped biphase symbols is shown schematically in figure 1. In concept each source bit gives rise to an odd impulse-pair, e(t), such that a logic 1 at source gives:

$$e(t) = \delta(t) - \delta(t - t_d/2) \qquad (1)$$

and a logic 0 at source gives:

$$e(t) = -\delta(t) + \delta(t - t_d/2) \qquad (2)$$

These impulse-pairs are then shaped by a filter $H_T(f)$, to give the required band-limited spectrum where:

$$H_T(f) = \begin{cases} \cos\dfrac{\pi f t_d}{4} & \text{if } 0 \le f \le 2/t_d \\ 0 & \text{if } f > 2/t_d \end{cases} \qquad (3)$$

and here

$$t_d = \frac{1}{1187.5}\ \text{s}$$

The data-spectrum shaping filtering has been split equally between the transmitter and receiver (to give optimum performance in the presence of random noise) so that, ideally, the data filtering at the receiver should be identical to that of the transmitter, i.e. as given above in equation (3). The overall data-channel spectrum shaping $H_o(f)$ would then be 100% cosine roll-off.

The specified transmitter and receiver low-pass filter responses, as defined in equation (3) are illustrated in figure 3, and the overall data-channel spectrum shaping is shown in figure 4.

The spectrum of the transmitted biphase-coded radio-data signal is shown in figure 5 and the time-function of a single biphase symbol (as transmitted) in figure 6.

The 57 kHz radio-data signal waveform at the output of the radio-data source equipment may be seen in the photograph of figure 7.

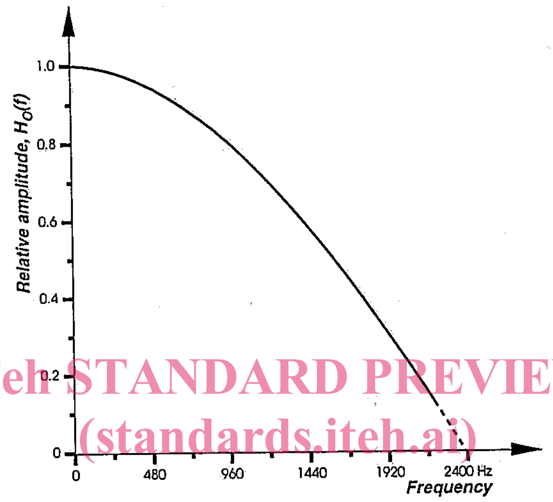


Figure 3: Amplitude response of the specified transmitter or receiver data-shaping filter

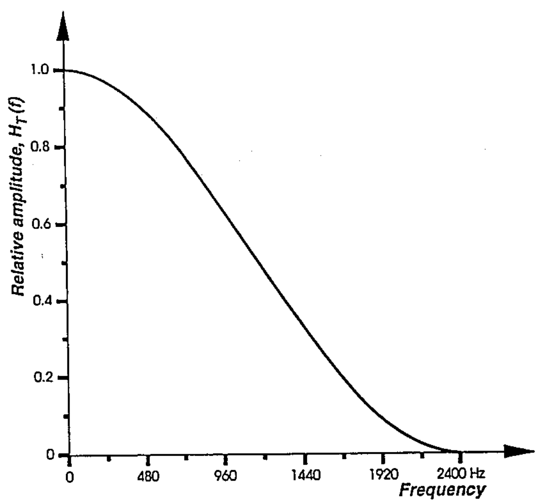


Figure 4: Amplitude response of the combined transmitter and receiver data-shaping filters

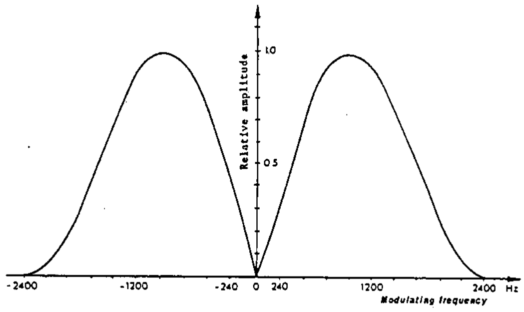


Figure 5: Spectrum of biphase coded radio-data signals

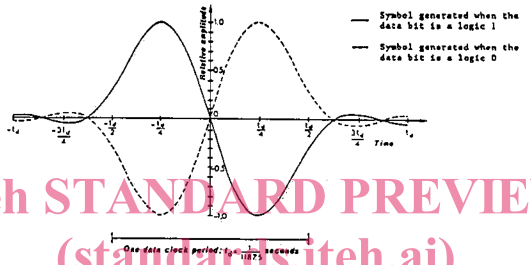


Figure 6: Time-function of a single biphase symbol

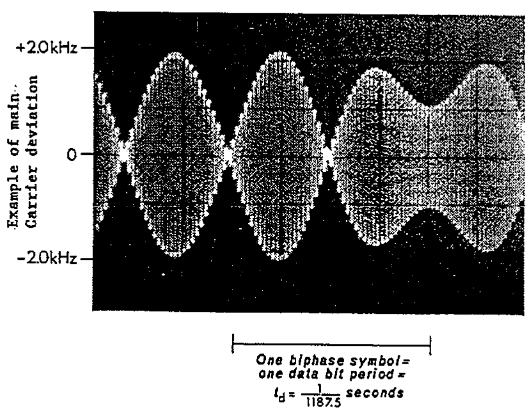


Figure 7: 57 kHz radio-data signals

2 Baseband coding (data-link layer)

2.1 Baseband coding structure

Figure 8 shows the structure of the baseband coding. The largest element in the structure is called a "group" of 104 bits each. Each group comprises 4 blocks of 26 bits each. Each block comprises an information word and a checkword. Each information word comprises 16 bits. Each checkword comprises 10 bits (see 2.3).

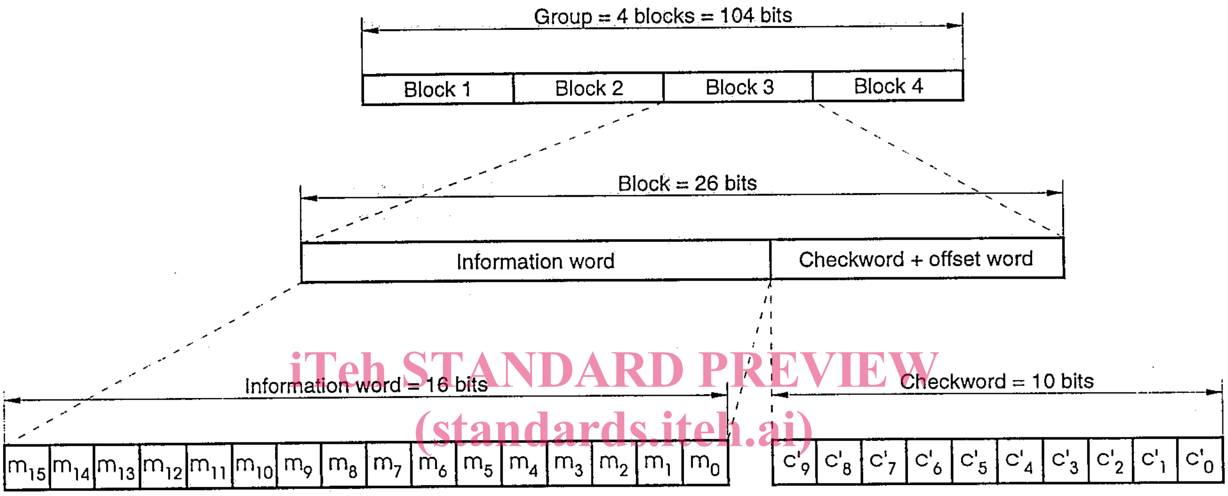
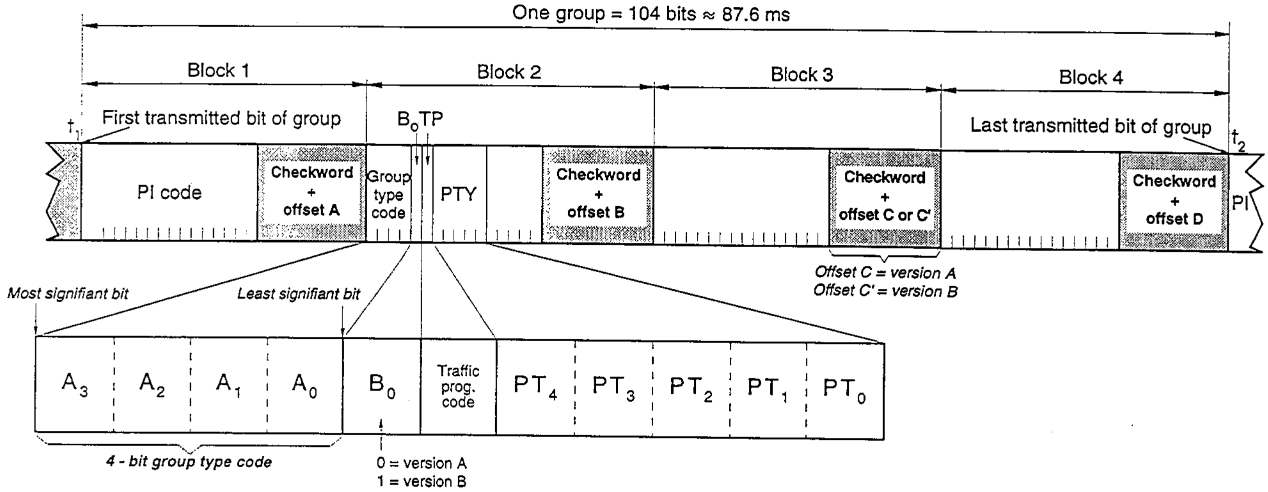


Figure 8: Structure of the baseband coding

2.2 Order of bit transmission

All information words, checkwords, binary numbers or binary address values have their most significant bit (m.s.b.) transmitted first (see figure 9). Thus the last bit transmitted in a binary number or address has weight 2^0.

The data transmission is fully synchronous and there are no gaps between the groups or blocks.



Notes to figure 9:

1. *Group type code = 4 bits (see 3.1)*
2. *B_o = version code = 1 bit (see 3.1)*
3. *PI code = Programme Identification code = 16 bits (see 3.2.1.1 and annex D)*
4. *TP = Traffic Programme Identification code = 1 bit (see 3.2.1.3)*
5. *PTY = Programme Type code = 5 bits (see 3.2.1.2 and annex F)*
6. *Checkword + offset "N" = 10 bits added to provide error protection and block and group synchronization information (see 2.3 and 2.4 and annexes A,B and C)*
7. *$t_1 < t_2$: Block 1 of any particular group is transmitted first and block 4 last*

Figure 9: Message format and addressing

2.3 Error protection

Each transmitted 26-bit block contains a 10-bit checkword which is primarily intended to enable the receiver/decoder to detect and correct errors which occur in transmission. This checkword (i.e. c'_9, c'_8, ... c'_0 in figure 8) is the sum (modulo 2) of:

a) the remainder after multiplication by x^{10} and then division (modulo 2) by the generator polynomial g(x), of the 16-bit information word,

b) a 10-bit binary string d(x), called the "offset word",

where the generator polynomial, g(x) is given by:

$$g(x) = x^{10} + x^8 + x^7 + x^5 + x^4 + x^3 + 1$$

and where the offset values, d(x), which are different for each block within a group (see 2.4) are given in annex A.

The purpose of adding the offset word is to provide a group and block synchronisation system in the receiver/decoder (see 2.4). Because the addition of the offset is reversible in the decoder the normal additive error-correcting and detecting properties of the basic code are unaffected.

The checkword thus generated is transmitted m.s.b. (i.e. the coefficient of c'_9 in the checkword) first and is transmitted at the end of the block which it protects.